Research Letter **A Low-Overhead Cooperative Retransmission Scheme for IR-UWB Networks**

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The UWB unique properties such as fine ranging and immunity to small scale fading are utilized in order to exploit the multiuser diversity in UWB networks. The optimal cooperation strategies in the absence of control packet overhead are analyzed in the proactive and reactive settings. It is shown that the proposed method achieves a considerable diversity gain while minimizing the overhead of control packet exchange that is required for coordination among the relays.

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1. Introduction

Due to the large bandwidth occupied by pulses, UWB signals are considered robust to small scale fading effects. In addition, UWB enables high accuracy ranging which can be used for the design of location-aware MAC and routing mechanisms. We exploit these properties of UWB, that is, availability of ranging information and immunity to small scale fading, in the design of an *U*WB-based *Co*operative *R*etransmission *S*cheme (*UCoRS*) [1].

Most of the existing distributed relay selection schemes, such as [2–4] rely on the Priority-Based Backoff Timer (*PBT*) mechanism to discover which relay is the best one at a time instance by sending a flag message. We note that like *PBT*, the other existing mechanisms such as *CMAC* [5] also require the exchange of the RTS/CTS and other control messages for every transmission. These cooperative methods may be inefficient for UWB networks. This is because the standard IR-UWB MAC protocol is ALOHA [6] and the exchange of RTS/CTS packets is not required prior to the data transmission in UWB. Furthermore, it is preferred to exchange fewer control packets due to the complex and costly UWB receiving procedure.

2. System Model

Figure 1 shows the system model. As can be seen, there are a source *S* and a destination *D*, and *N* relays R_i , $i =$

1, 2, *...* , *N*, in a slotted time domain, and each time slot consists of 2 subslots. At the transmission subslot (Tx), the source node sends data to its destination. At the Cx subslot, the relays retransmit the source data. In particular, the *i*th relay, *Ri*, decides to cooperate (i.e., retransmit the data) with probability *ai*.

Since accurate ranging information is available through UWB physical layer, we also presume that when *Ri* finds its distance to *S* and *D*, it broadcasts a packet to inform other nodes about these ranging information. Note that as long as the nodes do not move, the process of ranging and informing other nodes should be performed only once, which incurs much less overhead compared to sending control packets for *every* transmission.

The link success probabilities are denoted by *Pi* and *Qi*, as can be seen in Figure 1. The success probability of the *S-D* link is denoted by *P*₀. To calculate these values, we note that in time-hopping pulse position modulation, TH-PPM, the transmitted signal by node *i* is given by $s^i(t)$ = $\sum_{j=-\infty}^{\infty} \sqrt{E_p} \omega(t - jT_f - c_j^i T_c - \delta b_{\lfloor j/N_S \rfloor}^i)$, where *E_p* is the transmission energy per pulse, T_f and T_c are the frame and chip durations, $b^i_{\lfloor j/N_S \rfloor} \in \{0,1\}$ is the information bit to be sent, $\omega(t)$ is the monocycle pulse, and δ determines the time shift in the chip when the data bit is 1. Each frame consists of N_h chips, that is, $T_f = N_h T_c$. Moreover, each bit is repeated in N_S frames with different time hopping

codes, $c_j^i \in \{0, 1, \ldots, N_h - 1\}$, which results in additional (random) time shifts and hence increases the pulse immunity to interference.

The received signal from user *i* at node *j* is given by $[7]$ $r^{ij}(t) = \alpha^{ij} \sum_{c=1}^{C} \sum_{l=1}^{L} \beta_{cl} s^{i}(t - \tau_{cl}) + n(t)$, where $n(t)$ is AWGN with the power spectral density $N_0/2$, and α^{ij} denotes the *i*-*j* link gain. Since UWB pulses are robust to small scale fading effects, we consider only the channel pathloss, as defined in [1, 6]. Then, the bit error probability (BEP) in the absence of interference can be approximated by $P_{be}(d_{ij}) = (1/2) \text{erfc}(\sqrt{(\alpha^{ij} E_p N_S/2N_0)(1 - \rho(\delta))})$ [7], where $\rho(\delta)$ is the autocorrelation function of the monocycle pulse, $\omega(t)$. From the above-mentioned model, the probability that a packet with length *L* bits is successfully transmitted can be represented as follows:

$$
P_s(d) = 1 - (1 - P_{be}(d))^L.
$$
 (1)

This equation can be used to determine the values of *Pi* and *Qi* as a function of the relays' distances to *S* and *D*. Having obtained *Pi* and *Qi*s from (1), the next problem is to find the cooperation probabilities *ai* in order to maximize the *S*-*D* throughput.

3. Analysis

We assume that the packet level collision occurs if the signal strength of more than one packet is above the threshold at the receiver. Therefore, *D* successfully receives a useful data packet if either the *S*-*D* transmission in the Tx subslot is successful, or the transmission from one and only one of the relays in the Cx subslot is successful.

We consider two different settings, namely, the proactive and reactive modes. In the proactive mode, the decision is made prior to the source transmission. In the reactive mode, all relays listen for the data first and then decide to cooperate. Note that since message exchange between relays is not performed in *UCoRS*, a relay *Ri* is unable to find out the set of relays which have successfully decoded the packet from *S* at time slot *t*, denoted by *F*(*t*). In fact, the global optimum of the relay selection problem would be obtained if *F*(*t*) were available to the nodes.

In the proactive case, the expected success probability in a time slot is given by

$$
U(A) = P_0 + (1 - P_0) \sum_{i=1}^{N} \left[a_i P_i Q_i \prod_{j=1, j \neq i}^{N} (1 - a_j P_j Q_j) \right]. \tag{2}
$$

In order to find the optimal solution of (2), we use Lemma 1 in the Appendix. The following theorem gives the optimal solution.

Theorem 1. *Consider a cooperative network with one S-D pair and N relays. The optimal cooperation strategy to maximize the S-D throughput* $(U(A)$ *in* $(2))$ *is* $A^{(K)} = \{a_i = 1, i \leq 1\}$ *K*; $a_i = 0$, $i > K$ }, where *K* satisfies: $\sum_{i=1}^{K-1} (P_i Q_i / (1 - P_i Q_i))$ < 1, and $\sum_{i=1}^{K} (P_i Q_i/(1 - P_i Q_i)) \geq 1$, where relays are sorted in *descending order according to the values of* P_iQ_i *(i.e.,* $i \leq j \Leftrightarrow$

Figure 1: System model.

Figure 2: Comparison of packet delivery ratio (PDR) for different schemes.

 $P_iQ_i \geq P_jQ_j$ *); and* $A^{(K)}$ *denotes a binary vector whose first Kth elements are* 1*. (Proof is straightforward from Lemma 1.)*

Here, we mention that if $P_1Q_1 \geq 0.5$, then $K = 1$, and only R_1 will be active. In this special case, the result is in agreement with [3]. The reactive and global optimum cooperation strategies can be derived using the same reasoning as Theorem 1, as discussed in detail in [1].

4. Performance Evaluation

Figure 2 compares the packet delivery ratio (PDR) for different scenarios. As can be seen, the proactive performance is near to the maximum achievable throughput. Furthermore, as expected, both reactive and proactive methods outperform the noncooperative case.

Figure 3(a) shows the effect of increasing the number of relays on the achieved PDR in *UCoRS* for different *S*-*D* link qualities. As can be seen, adding one relay can significantly increase the PDR of the direct link. However, the achieved PDR in *UCoRS* is upper bounded by a function of d_{SD} , regardless of number of available relays.

Figure 3(b) shows the asymptotic achievable throughput of *UCoRS*, *PBT*, and noncooperative schemes as a function of P_0 when $N \rightarrow \infty$. As stated previously, the throughput advantage of *PBT* over *UCoRS* is at the expense of control packet exchange for every data transmission, which may not be efficient in UWB. More details can be found in [1].

Figure 3: Simulation results.

5. Conclusion

We introduced *UCoRS*, a simple *U*WB-based *Co*operative *R*etransmission *S*cheme, that utilizes the unique properties of IR-UWB technology for achieving multiuser diversity in UWB in the proactive and reactive settings. The amount of control packet overhead is minimized in *UCoRS* in order to eliminate the corresponding energy cost at the UWB receivers.

Appendix

Lemma 1. Assume a set of variables $Z = \{z_i\}, i = 1, 2, \ldots, n$, *that can take on real values between* 0 *and* $1 > m_1 \ge m_2 \ge$ $\sum_{i=1}^{n}(z_i)\prod_{j\neq i}(1-z_j)$ *is obtained when* $z_i = m_i$ *, i* ≤ *K, and* $\cdots \ge m_n$ *, respectively. Then, the maximum value of* $X(Z) =$ $z_i = 0, i > K$ *, where K satisfies* $\sum_{i=1}^{K} (m_i/(1 - m_i)) \geq 1$ *, and* $\sum_{i=1}^{K-1} (m_i/(1 - m_i)) < 1$.

Proof. Taking the partial derivative of *X*(*Z*), we have

$$
\frac{\partial X(Z)}{\partial z_i} = \prod_{j \neq i} (1 - z_j) - \sum_{j \neq i} \left(z_j \prod_{k \neq i,j} (1 - z_k) \right)
$$
\n
$$
= \prod_{j \neq i} (1 - z_j) \left(1 - \sum_{j \neq i} \frac{z_j}{1 - z_j} \right). \tag{A.1}
$$

Therefore, $\partial X(Z)/\partial z_i > 0 \Leftrightarrow \sum_{j=1, j \neq i}^{n} (z_j/(1-z_j))$ < 1, and *∂X*(*Z*)*/∂zi > ∂X*(*Z*)*/∂zj* ⇔ *zi > zj*. According to these two results, in order to maximize *X*(*Z*), the *K* "best" variables (with looser bounds) should be set to their maximum values and other variables should be set to 0. The required conditions on *K*are also clearly observed from the above-mentioned equations. Note that if $m_1 \geq 0.5$, then $K = 1$. \Box

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